

The Time Programmable Frequency Comb: Generation and Application to Quantum-Limited Dual-Comb Ranging

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The classic self-referenced frequency comb acts as an unrivaled ruler for precision optical metrology in both time and frequency^{1,2}. Two decades after its invention, the frequency comb is now used in numerous sensing applications³⁻⁵. Many of these applications, however, are limited by the tradeoffs inherent in the rigidity of the comb output and operate far from quantum-limited sensitivity. Here we demonstrate an agile programmable frequency comb where the pulse time and phase are digitally controlled with ± 2 attosecond accuracy. This agility enables quantum-limited sensitivity in sensing applications since the programmable comb can be configured to coherently track weak returning pulse trains at the shot-noise limit. To highlight its capabilities, we use this programmable comb in a ranging system, reducing the required power to reach a given precision by $\sim 5,000$ -fold compared to a conventional dual-comb system. This enables ranging at a mean photon per pulse number of $1/77$ while retaining the full accuracy and precision of a rigid frequency comb. Beyond ranging and imaging⁶⁻¹², applications in time/frequency metrology^{1,2,5,13-23}, comb-based spectroscopy²⁴⁻³², pump-probe experiments³³, and compressive sensing^{34,35} should benefit from coherent control of the comb-pulse time and phase.

As applications of frequency combs have expanded, their uses have extended beyond functioning simply as a reference ruler³⁻⁵. For example, many experiments combine two or more frequency combs for active sensing including precision ranging and imaging⁶⁻¹², linear and non-linear spectroscopy²⁴⁻³², and time transfer^{13-20,23}. In these applications, the multiple fixed combs serve as differential rulers by phase-locking them to have a vernier-like offset between their frequency comb lines, or their pulses in time. While these applications exploit the accuracy and precision of frequency combs, they operate nowhere near the quantum (or shot noise) limit, despite the use of heterodyne detection, because of effective dead time due to sensing the incoming signal-comb light via a comb with a deliberately mismatched repetition frequency. Consequently, there are strong tradeoffs in measurement speed, sensitivity and resolution^{24,36,37}. In some dual-comb ranging and spectroscopy demonstrations, these penalties have been partially addressed by incoherent modulation of the comb³⁸⁻⁴¹ but not eliminated.

Here, we overlay a self-referenced optical frequency comb with synchronous digital electronics for real-time coherent control of the comb's pulse train output. We manipulate the frequency comb's two phase locks to dynamically control and track the time and phase of the frequency comb's output pulses at will. The temporal placement of the comb pulses is set with ± 2 attoseconds accuracy with a range limited only by slew rate considerations. This time programmable frequency comb (TPFC) goes beyond the "mechanical gear box" analogy often applied to optically self-referenced combs⁵, replacing it with a digitally controllable, agile, coherent optical pulse source. The agility of the TPFC enables many more measurement modalities than a rigid frequency comb. In sensing applications, the TPFC can enable quantum-limited detection with the full accuracy and precision of the frequency comb, avoiding the penalties discussed previously. To achieve these combined advantages, the TPFC is configured as a tracking optical oscillator in time and phase so that it effectively locks onto an incoming weak signal pulse train for coherent signal integration.

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50 As an immediate example, we incorporate the TPFC into a dual-comb ranging system. The result
51 is quantum-limited sensing that sacrifices none of the exquisite accuracy and precision of
52 frequency-comb measurements. Here, we show a precision floor of 0.7 nm (4.8 attoseconds in
53 time-of-flight) in ranging, which exceeds previous conventional dual-comb ranging
54 demonstrations^{6-8,42-44}. In addition, the tracking dual-comb ranging detects a weak reflected signal-
55 comb pulse-train with a mean photon number per pulse of only 1/77 at a sensitivity within a factor
56 of two of the quantum limit. Detection of signals at even lower mean photon per pulse numbers is
57 possible by reducing the measurement bandwidth. In contrast, conventional dual-comb ranging
58 would require a return signal 37 dB or 5000x stronger to reach the same level of performance.

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60 The uses of the TPFC go well beyond acting as a tracking optical oscillator. It should enable many
61 more time-based measurement schemes than the conventional vernier approaches using fixed
62 frequency combs. For example, in multi-comb sensing, the relative time offset between the
63 frequency combs can be adjusted so as to mimic a higher repetition rate system while retaining the
64 benefits of a lower repetition rate system, *e.g.* higher pulse energy and tight stabilization. Arbitrary
65 patterns can enable future compressive sampling³⁵. In time/frequency metrology, the comb can
66 provide accurately adjustable timing signals, modulation capabilities for noise suppression, and
67 optically-based time interval standards⁴⁵. Multiple TPFCs could be used for pump-probe
68 experiments with digital control of pulse spacing replacing delay lines or chirp-induced delays³³.

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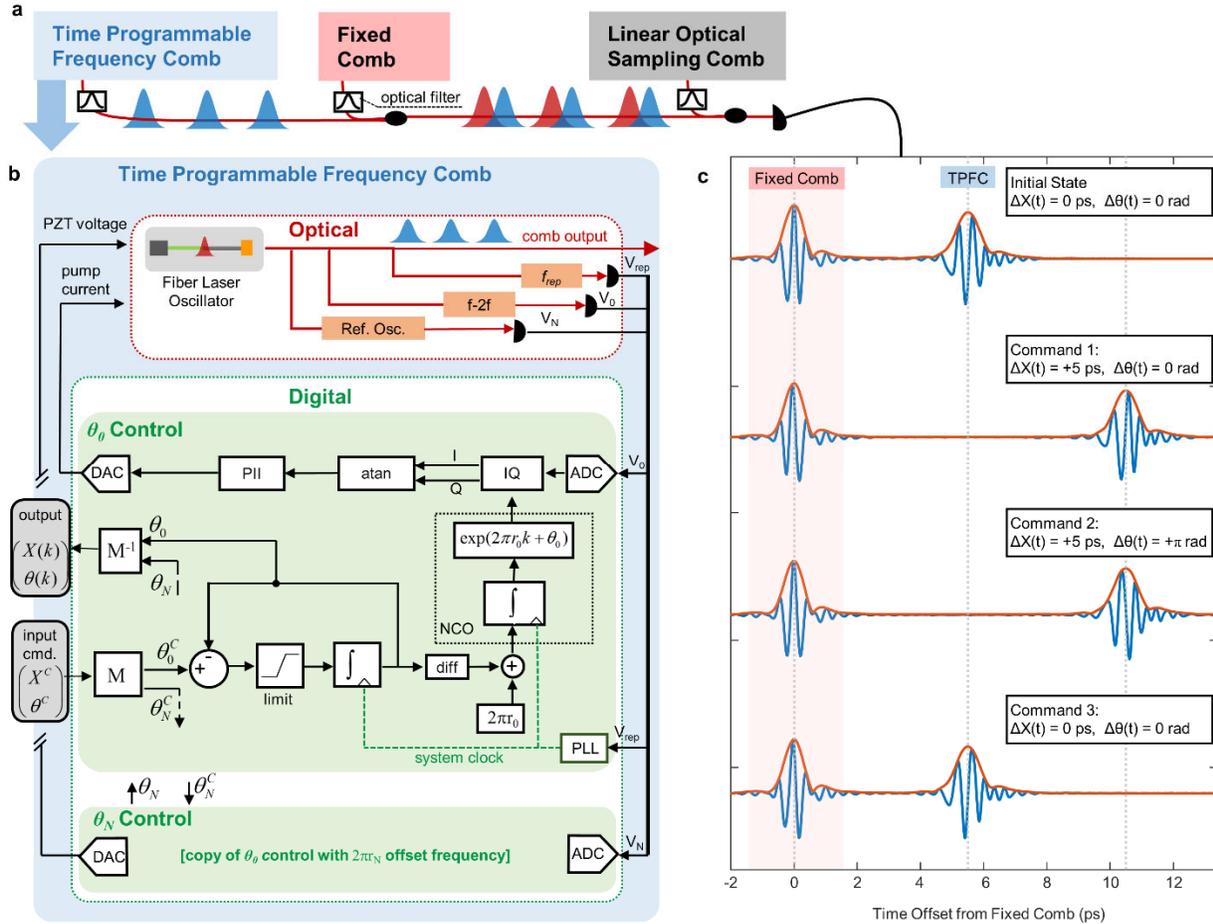
70 In this article, we first describe the TPFC and its capabilities generally. We then explore a specific
71 application by integrating the TPFC into a dual-comb ranging system. Finally, we discuss the
72 potential benefits of a TPFC in comb-based sensing more generally, including in LIDAR,
73 spectroscopy, and time transfer.

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75 **RESULTS**

76 **Generation of a Time Programmable Frequency Comb**

77 The TPFC requires two parts: an *optically* self-referenced frequency comb and the electronics to
78 track and control the time and phase of the comb pulses. (See Methods Eqn. 3 for a definition of
79 the time and phase of the comb pulses.) While the electronic system need not be exclusively
80 digital, it does need to track the programmed comb time and phase at the attosecond level over
81 long (hours to weeks) durations. Here, we use a fixed-point number whose least significant bit
82 corresponds to < 1 attosecond shift in time. When combined with an integer pulse number in an
83 80-bit number, the pulse timing can be specified with zero loss of accuracy for over 1 week at 1-
84 as precision, thereby providing well beyond 10^{19} -level control of the comb timing,
85 commensurate with next-generation optical clocks. As for the comb, any self-referenced comb
86 could be converted into a TPFC; here, we generate a TPFC using a fiber-based comb.



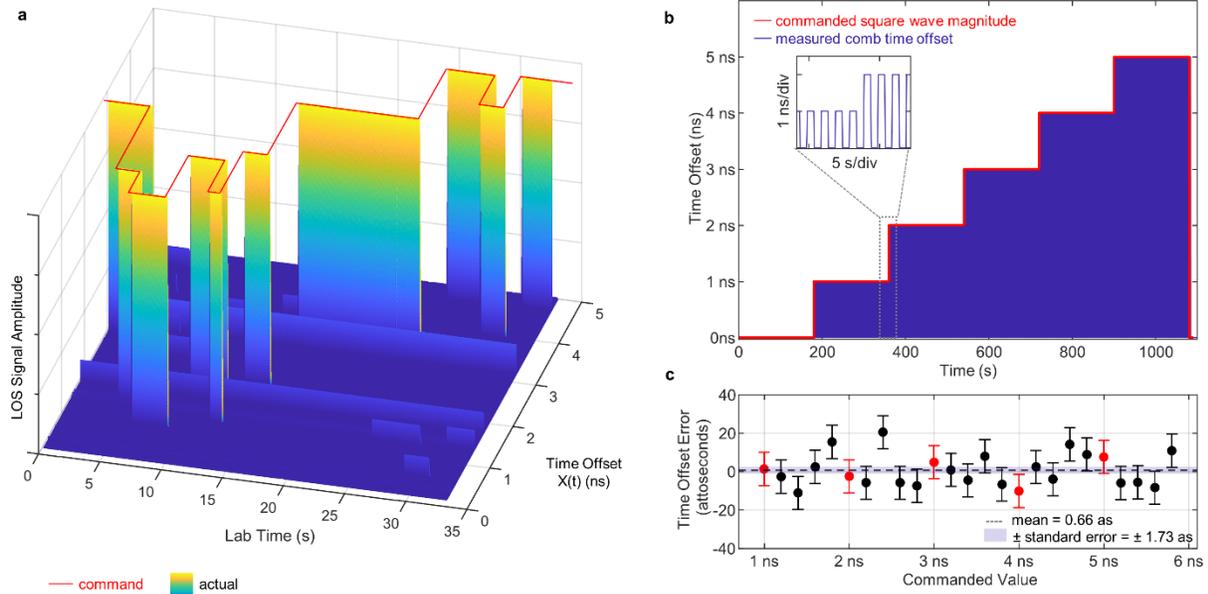
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88 **Figure 1: A time programmable frequency comb (TPFC).** (a) The TPFC output is
89 measured with respect to a second fixed frequency comb through linear optical sampling
90 (LOS) against a third frequency comb with an offset repetition frequency. The frequency
91 combs operate at $f_{rep} \sim 200$ MHz with a 5-ns pulse spacing. All pulses are spectrally filtered
92 to a Gaussian 10.1-nm wide shape, corresponding to 355 fs pulse duration (See
93 Methods). (b) Schematic of the TPFC. A self-referenced Er:fiber frequency comb is
94 controlled with digital electronics clocked off the detected comb repetition rate signal (V_{rep}).
95 The digital section receives the carrier-envelope offset signal (V_0), the optical beat signal
96 (V_N), along the comb pulse timing and phase commands, X^C and θ^C , which are combined
97 to give the control phases θ_0^C and θ_N^C through the (trivial) matrix M . These are passed to
98 their respective digital control loop (see Methods). The control efforts for θ_0^C and θ_N^C
99 adjust the phase-locked loops (PLLs) controlling the comb's two degrees of freedom. The
100 system tracks the actual phases, θ_0 and θ_N as fixed point numbers, which are combined to
101 give the actual pulse timing and phase, $X(k)$ and $\theta(k)$, for every comb pulse number k .
102 IQ: in-phase/quadrature demodulator, PII: proportional-integral-integral controller,
103 NCO: numerically controlled oscillator, r_0 and r_N : offset frequencies of the phase locks in
104 units of f_{rep} (see Methods). (c) LOS (blue trace) and their envelopes (red trace) for the fixed
105 comb (at $X=0$) and the TPFC at the given (X, θ) values. The LOS magnification of the time
106 axis is 10^6 .

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Figures 1 and 2 describe the TPFC and its output characterization. In a self-referenced comb, phase-locked loops (PLLs) stabilize the frequency of the N^{th} comb tooth, f_N , with respect to a CW reference laser, and the frequency of the 0^{th} comb tooth, f_0 (the carrier-envelope offset frequency). The PLL locks both frequencies to a known fraction of f_{rep} , which is self-referentially defined as $f_{rep} \equiv (f_N - f_0)/N$ ^{1,2,4,5}. These PLLs also set the phases of the N^{th} and 0^{th} comb tooth frequencies, θ_N and θ_0 , to arbitrary but fixed values. Here, we manipulate these phases to control both the comb-pulse phase, θ , and the comb-pulse time offset which is given by $X = (\theta_0 - \theta_N)/(2\pi N f_{rep})$ in direct analogy to f_{rep} 's definition above. The digital control exploits the optical frequency division of N inherent to optically self-referenced combs since a single 2π shift in the phase of either PLL leads to a time shift ~ 5 femtoseconds. The TPFC outputs both a train of optical pulses and the corresponding synchronous digital values of pulse time, X , and pulse phase, θ (Fig. 1b).

The TPFC is both agile and accurate (Figure 1c and Figure 2); the output time of a comb pulse can be adjusted arbitrarily. Yet at any instant, we know exactly, to fractions of an optical cycle, by how much the output time (and phase) has been shifted. For rapid changes in the TPFC output, the settling time of the PLLs can be taken into account either via modeling or by including the digital phase error signal from the two PLLs. It is the exactness of the performed step relative to the commanded step (Fig. 2b) and the ability to control the steps in real time that stand in contrast to earlier work. Shown in Fig. 2b, the accuracy of the timing control, X , with respect to the underlying CW reference laser is 0.66 ± 1.73 attoseconds.

Here the maximum slew rate between time steps was conservatively set to 40 ns/s to eliminate the possibility of cycle slips in the PLL during motion. The use of an input tracking filter for the PLL signals should enable slew rates as high as 1 $\mu\text{s/s}$, limited only by the actuators. (See Methods).



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Figure 2: Illustration and characterization of the time programmability of the TPFC through LOS, using the setup in Fig 1a. (a) The TPFC pulse train, presented as a surface plot, where each slice in lab time represents a complete LOS measurement as in Figure 1c. The TPFC pulse is located at the LOS signal peak and follows the commanded arbitrary step pattern (red line). Multiple reflections within the setup appear as small satellite pulses. (b) Repeated stepping of the TPFC timing to verify accuracy. Steps are performed at 1 Hz, measured by LOS at 6 kHz (blue line), and the commanded step size (red line) is changed every 3 minutes. The 1 Hz modulation allows accurately measuring the step size by removing fiber optic path length drifts. (c) The error between the actual and commanded pulse times for the data in 2b (red circles). Each point is a 3-minute average over $\sim 1\text{M}$ individual LOS measurements. This measurement was repeated for multiple different commanded time steps (black circles). The uncertainty bars are based on the LOS measurement noise and residual comb timing jitter. The average difference is 0.66 attoseconds ± 1.73 attoseconds (standard error). There is no observed reduction in accuracy or precision despite moving the TPFC over the full 5 ns non-ambiguity range.

Example Application: Dual-Comb Ranging with a TPFC

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To demonstrate the advantages of the TPFC in dual-comb sensing, we consider ranging⁶⁻⁸. In dual-comb ranging, pulses with bandwidth τ_p^{-1} from a comb are reflected off an object, and their time-of-flight is detected by heterodyning them against a second comb. This measurement has a resolution of $\Delta R = c\tau_p / 2$, which characterizes the ability to distinguish two adjacent reflections. It has a non-ambiguity range $R_{NA} = c / (2f_{rep})$, associated with “which pulse” is detected. (This ambiguity can be removed by changing f_{rep} and repeating the measurement⁷). The accuracy is set by the comb’s reference oscillator or knowledge of the index of refraction. In certain applications, absolute calibration and instability of the reference plane will also factor into the accuracy. The precision is the uncertainty in the peak location of the reflection. At best, the precision is equal to the resolution divided by the signal-to-noise ratio (SNR):

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$$\sigma_R = \frac{C}{2\ln(2)} \frac{\Delta R}{SNR_S} \quad (1)$$

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where the $(2\ln(2))^{-1}$ factor arises from assuming Gaussian pulses (see Methods). The shot-noise limited signal-to-noise ratio, $SNR_S = \sqrt{\eta n_s}$ where η is the detector quantum efficiency and

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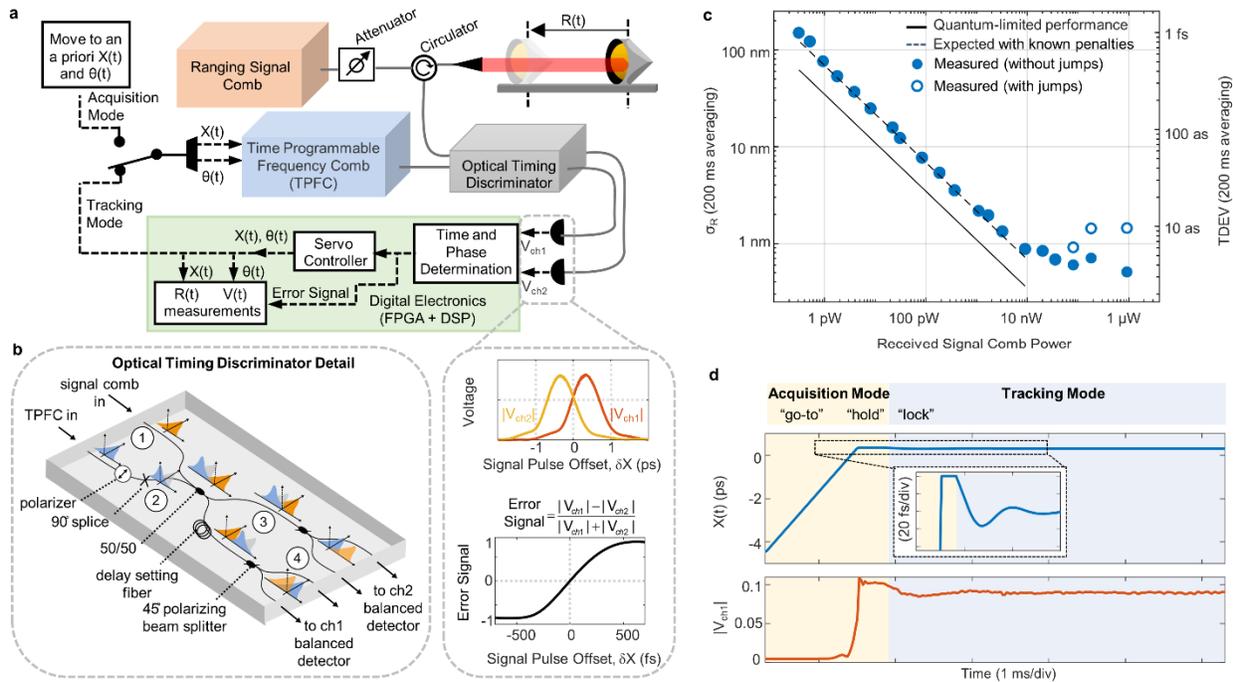
$n_s = P_{rec} T / (h\nu)$ is the number of signal photons for a received power P_{rec} and integration time,

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T . The constant C quantifies how far the precision is from the quantum limit. It can be related to the power penalty as $P_p = C^2$. An optimal quantum-limited ranging system operates at

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$$C = P_p = 1.$$



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Figure 3: Dual-comb ranging with a time programmable frequency comb. (a) System

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diagram. The TPFC can be run in two modes, acquisition or tracking, as described in the

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main text. (b) The timing discriminator is a dual Mach-Zehnder interferometer

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constructed with polarization-maintaining fiber optics in which (1) both comb pulses enter

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with the same polarization and (2) the TPFC pulse is rotated to the fast-axis. Then (3) the

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pulses are mixed, a delay between pulses is added to one arm and (4) the pulses are

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projected back into the same polarization for balanced heterodyne detection to yield the

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two output signals V_{ch1} and V_{ch2} . These signals are demodulated and their magnitudes

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combined to generate a power-insensitive error signal of the differential comb pulse time

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offset, while their phase yields the differential comb pulse phase. (c) Range precision

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(deviation) σ_R (left axis) and corresponding time deviation, TDEV (right axis) at 200-ms

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averaging time versus the signal-comb power at the balanced photodetectors. The

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measured precision follows the quantum limit (Eq. (1)) from 0.33 ± 0.03 pW to $10 \pm$

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1.0 nW with a penalty of $C = 2.16$, reaching a systematic noise floor below 1 nm (7 as),

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which is 2-10x below previous dual-comb ranging experiments^{6-8,42-44} and attributed to

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residual fiber path-length fluctuations. At the three high power points, the deviation is

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given both for the full data trace and for a subset of the data without ~ 6 sudden

191 **ranging/timing jumps of <50-nm amplitude which are attributed to nonlinear optical**
192 **feedback (and only appear at these higher powers). (d) Example handover between**
193 **acquisition and tracking modes. Here the TPFC was commanded to move from $X(t) = -5$**
194 **ps to $X(t) = 0$ ps where it acquires the signal pulse.**
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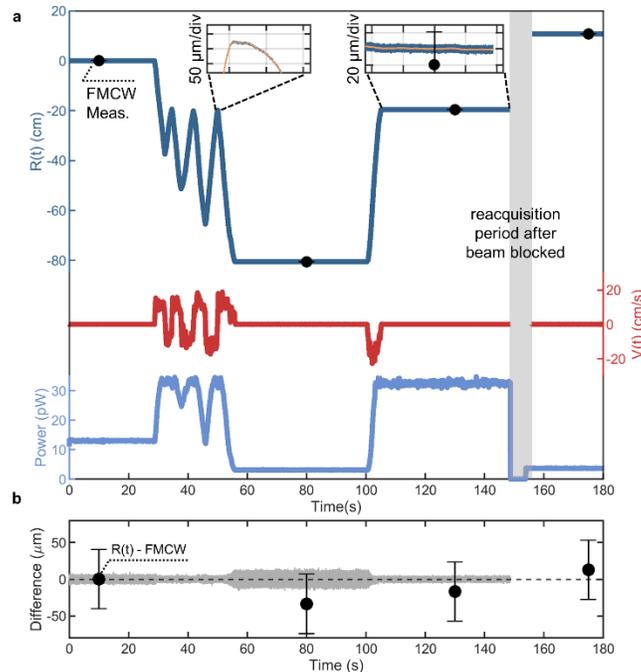
196 Conventional dual-comb ranging operates at a high power penalty and with significant tradeoffs.
197 In these systems, the second comb's repetition rate is offset by Δf_r to serve as a linear sampling
198 comb. It repeatedly scans the entire non-ambiguity range, R_{NA} , at a measurement rate
199 $T^{-1} = \Delta f_r \leq f_{rep} \Delta R / (4R_{NA})$. The inherent tradeoffs in T , R_{NA} , and ΔR have led to dual-comb
200 ranging implementations using very different frequency combs and covering three orders of
201 magnitude in T and ΔR but all facing these strong constraints^{6-8,42-44}. Moreover, in all cases, the
202 power penalty $P_p \approx \Delta R / R_{NA} = f_r \tau_p$ is severe, ranging from 14 dB to 38 dB^{7,42}, because of the
203 repeated scanning of the entire non-ambiguity range.
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205 Here, as shown in Figure 3, we replace the sampling frequency comb by the TPFC to overcome
206 these tradeoffs and stiff power penalty. A coherent timing discriminator, shown in Fig. 3b,
207 measures the relative time and phase between the TPFC and signal comb pulses. Creating two
208 time-shifted copies effectively shapes the TPFC pulse to optimize detection of the time of the
209 incoming signal-comb pulses^{46,47}. The two heterodyne signals output by the timing discriminator
210 have a nominal carrier frequency of 10 MHz, set by the relative phase locks of the two combs.
211 The signals are then bandpass filtered and demodulated. The choice of bandwidth is a tradeoff
212 between update rate and sensitivity. Here, we use 26-kHz to be well above typical ~kHz
213 mechanical vibrations while allowing for a fairly long 13 microsecond coherent integration
214 time. The amplitudes of the two demodulated signals are combined to compute the time offset
215 between the TPFC and signal comb (see Fig. 3b) while their phase yields the differential carrier
216 phase of the TPFC and signal comb pulses whose derivative is the Doppler shift. Critically, the
217 combination of the two channels from the timing discriminator gives a time offset measurement
218 that is independent of the incoming signal power.
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220 This system runs in two modes: acquisition mode and tracking mode, both of which differ from
221 conventional dual comb ranging. In acquisition mode, $X(t)$ is scanned until the tracking comb's
222 timing matches the incoming signal pulse train. While it is possible to scan the entire non-
223 ambiguity range, we can also make use of *a priori* information to scan the TPFC's over much
224 less than the non-ambiguity range. The information could be provided from external sources or
225 from a Kalman filter if previous range/Doppler measurements are available. Once the system
226 acquires the appropriate reflection, it switches to tracking mode (Fig. 3d). Tracking mode
227 implements a pulse-timing lock and a carrier-frequency lock based on the timing discriminator
228 outputs (see Methods). The combination of the control and error signals from the time and
229 frequency locks in turn yield the range and Doppler velocity of the target object.
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231 In tracking mode, the ranging precision nearly reaches the quantum limit of Eq. (1) (Fig. 3c).
232 This nearly quantum-limited precision ranging is demonstrated at a rapid 26-kHz measurement
233 rate with as little as 0.33 ± 0.03 pW of return power ($SNR_S = 9.5$), which corresponds to only
234 $1/77$ mean photons per pulse. There is a slight penalty of $C = 2.16x$ due primarily to differential
235 dispersion between the comb pulses (see Methods). With additional optimization, C could be

236 reduced to 1, and with squeezing, to <1 ⁴⁶. With these same 200-MHz combs, conventional dual-
 237 comb ranging would suffer from a power penalty $P_p = 37$ dB ($C = 71$). Finally, momentary loss
 238 of signal is not an issue. If brief enough that the object position does not differ from prediction,
 239 e.g. from a Kalman filter, by more than $\sim \pm 2\Delta R = \sim \pm 100$ μm , tracking simply resumes. For
 240 even a 10g acceleration, a ~ 1.5 ms duration loss of signal is tolerable. Sometimes the signal loss
 241 may be too long, for example in strong air turbulence⁴⁸, in which case the system can transition
 242 to acquisition mode using previous data to limit the scan, as discussed above.
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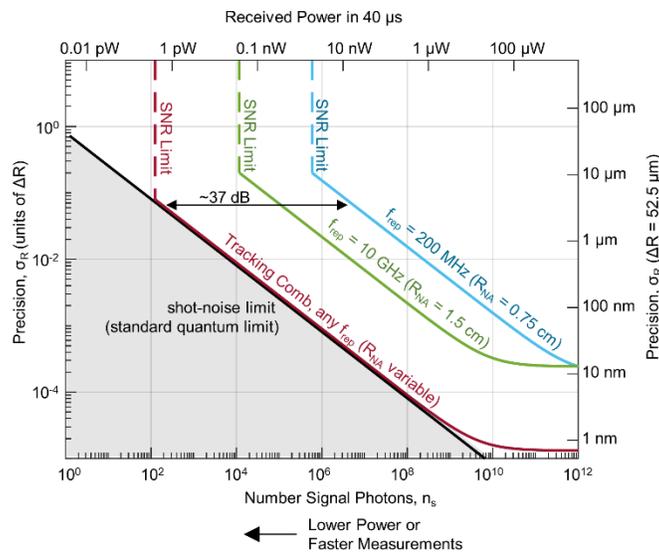
244 **Figure 4: Ranging and velocity data to a moving retroreflector. (a) The range (top left axis,**
 245 **dark blue trace) is measured from the summed control and error signals for $X(t)$ in**
 246 **tracking mode at a 26 kHz rate. The velocity (right middle axis, red trace) is calculated**
 247 **from summed control and error signals for $d\theta(t)/dt$. At 150 seconds, the beam was blocked**
 248 **and the target moved, triggering a re-acquisition. Before and after re-acquisition the**
 249 **measured range agrees with a commercial FMCW ranging system (black circles), to within**
 250 **the FMCW uncertainty driven by target vibrations (see Methods). The relative range also**
 251 **agrees well with the interferometric range from the unwrapped carrier phase (yellow trace**
 252 **in the two insets), after applying an overall offset. (b) Difference between the dual-comb**
 253 **range and FMCW range (black circles) and between the dual-comb range and**
 254 **interferometric range from the unwrapped carrier phase for the initial section with**
 255 **continuous signal (grey trace). The range deviation of this latter difference reaches 10 nm**
 256 **and 5 nm at 10-s averaging for the time periods with received powers of 3.2 ± 0.3 and $32 \pm$**
 257 **3.0 pW, respectively. (See Extended Data Figure 1.)**
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260 Figure 4 shows range data taken while moving the rail-mounted retroreflector. The dual comb
 261 system tracks the retroreflector as it reverses direction at velocities up to 20 cm/s. The signal is
 262 blocked at 150 s and the retroreflector moved, after which the absolute range was reacquired by
 263 scanning over a ± 37.5 cm window. For validation, we compare to a commercial frequency-
 264 modulated continuous-wave (FMCW) system at a few static rail positions after calibrating out

265 differential range offsets. The two agree to within the FMCW measurement uncertainty of ± 40
 266 μm due to target vibrations amplified by the FMCW's intrinsic range-doppler coupling
 267 (Extended Data Figure 2). Finally, the coherent timing discriminator also outputs the relative
 268 carrier phase between the signal comb and TPFC, whose derivative yielded the velocity above.
 269 This phase can also be unwrapped to provide relative range during periods of continuous signal
 270 (Fig. 4, yellow trace) as in Ref. ⁷ and similar to CW interferometry (except avoiding systematic
 271 errors from spurious reflections). This unwrapped carrier phase agrees with the tracking range to
 272 a precision limited by the tracking range noise which follows Eq. (1), after accounting for a
 273 $\sim 1.5\times$ chirp-induced penalty in C from the fiber optic path to reach the rail system.
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275 DISCUSSION

276 A number of existing or potential applications should benefit from the abilities illustrated in Fig.
 277 1-3, specifically to (1) set the time and phase of the comb's output pulses, (2) coherently scan the
 278 relative temporal spacing between two frequency combs over a specified limited range rather
 279 than the full inverse repetition rate, thereby mimicking a higher repetition-rate comb while
 280 avoiding limitations of lower pulse energy, and (3) operate as a precision optical tracking
 281 oscillator in time and frequency for shot-noise limited sensing. Below we discuss three different
 282 general application areas: LIDAR, time metrology, and spectroscopic sensing.



283
 284 **Figure 5: Precision versus received signal photons, $n_s = SNR_S^2 / \eta$, for conventional dual-**
 285 **comb ranging at $f_{rep} = 200$ MHz (blue line) and $f_{rep} = 10$ GHz (green line), compared to tracking**
 286 **dual-comb ranging (red line) and the standard quantum limit for heterodyne detection**
 287 **(black line). The left/bottom axes are in normalized units while the right/top axes are scaled**
 288 **to values similar to those used in this work. For conventional dual-comb ranging, the scaling**
 289 **depends strongly on f_{rep} and the precision is always much worse than the standard quantum**
 290 **limit. The tracking dual-comb ranging (red line) is independent of f_{rep} and can reach the**
 291 **standard quantum limit, though here shown with a 10% penalty. We assume a system noise**
 292 **floor of 0.7 nm, taken from this work, for the tracking comb, and ~ 10 nm for the conventional**
 293 **dual-comb systems^{7,42}. For all three ranging configurations, we require a minimum detection**
 294 **SNR of ~ 10 for reasonable detection statistics, as indicated by the SNR limit. The 37 dB**
 295 **improvement, however, is independent of this chosen limit.**

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297 As already discussed, frequency combs have a natural connection to precision LIDAR. Figure 5
298 and Extended Data Table 1 together compare conventional dual-comb ranging^{7,42-44}, tracking
299 dual-comb ranging and FMCW ranging⁴⁹, which is the standard approach to high-resolution
300 optical ranging. For all three, the resolution is set by the optical bandwidth and the accuracy by
301 the comb referencing or knowledge of the index of refraction of air. (Comb-assisted FMCW
302 ranging can transfer frequency-comb accuracy to FMCW LIDAR⁵⁰.) Both tracking dual-comb
303 ranging and FMCW ranging can reach the shot-noise limit and exploit the optical carrier phase.
304 However, the update rate of FMCW ranging is limited by the laser sweep time and its
305 uncertainty can be degraded by target vibrations. Tracking dual-comb ranging avoids vibration-
306 related systematics. It could provide range-resolved vibrometry in a cluttered environment or
307 surface imaging through turbulent media even with speckle-induced dropouts and high loss. (A
308 10-mW launch power will still provide sufficient 1-pW return power to enable a 26-kHz
309 measurement rate given a -100 dB reflection). More generally, there are strong overlaps with
310 conventional RF pulse-Doppler radar and the tracking comb could therefore have interesting
311 applications to high-bandwidth synthetic aperture LIDAR⁵¹.

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313 In time-frequency metrology, a TPFC phase-locked to an optical atomic clock provides an
314 optical timescale with the ability to “adjust” its output time to synchronize with other signals. A
315 TPFC could also enable calibration of time interval counters⁴⁵. The time-interval standard would
316 follow Fig. 1c, where the TPFC allows precisely defined variable pulse spacing from
317 nanoseconds to femtoseconds. This offers the prospect of a time interval standard that spans 6
318 orders of magnitude with attosecond precision and an accuracy directly tied to a secondary
319 representation of the second. For absolute optical frequency measurements, the TPFC also
320 enables determination of the mode number N by applying a shift $\Delta\theta_0 = 2\pi N$, which will lead to
321 a time shift of exactly $\Delta X = f_{rep}^{-1}$ for the correct N . Any integer error in N appears as a multiple of
322 5-fs offset in time, which can be resolved with a second comb and a timing discriminator. In
323 secure optical communications, the programmable comb might enable quadrature pulse phase-
324 position modulation if implemented with high-speed actuators. Finally, the TPFC has interesting
325 applications to comb-based long-distance free-space time transfer^{13-17,19,20,23,48} as it provides
326 similar advantages as in dual-comb ranging⁵².

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328 The TPFC can also break tradeoffs that limit comb-based linear and non-linear spectroscopy.
329 Relatively low repetition-rate frequency combs (100-MHz to 1-GHz) provide the high pulse
330 energies needed for nonlinear spectroscopy or for spectrally broadening over the desired spectral
331 band⁵³⁻⁵⁵. However, the spectral resolution set by these low repetition rates is often poorly
332 matched to the application leading to significant “deadtime” or SNR reduction in multi-comb-
333 based spectroscopy^{24,26-31,54,55}. The TPFC can circumvent this problem by coherently scanning
334 over a limited time offset between two or more frequency combs, as first demonstrated
335 incoherently in the early dual-comb spectroscopic work of Schliesser *et al.*³⁸. In this way, a low-
336 repetition frequency TPFC can act as a frequency comb with an effectively much higher
337 repetition rate, set by the inverse of the temporal scan window, at shot-noise limited sensitivity.
338 Going further, the ability to jump the frequency comb pulse phase and timing could enable
339 compressive sampling in dual-comb or multi-comb sensing applications with a concomitant
340 increase in measurement rate. Recent modeling³⁵ and preliminary experiments³⁴ have highlighted
341 the advantages of this dynamic control for dual-comb spectroscopy. Finally, in nonlinear

342 spectroscopy, temporal control could enable time-ordered multi-photon excitation, following the
343 comb-based spectroscopy of Rubidium³³ but with programmable control.

344

345 **CONCLUSION**

346 The time programmable frequency comb combines the precision and accuracy of a self-
347 referenced frequency comb with flexibility in time and phase and 2-attosecond accuracy. Here,
348 the TPFC is based on a fiber frequency comb, but any self-referenced comb (or comb locked to
349 widely separated optical oscillators) with control electronics capable of tracking and
350 manipulating phase could act as a TPFC. Through a dual-comb ranging demonstration, we show
351 the TPFC can operate as an optical tracking oscillator in time and frequency, yielding nearly
352 quantum-noise limited ranging with 0.7 nm precision. Finally, dual-comb ranging is just one
353 application and the TPFC has equal promise in relaxing tradeoffs with repetition frequency and
354 improving SNR in other multi-comb sensing and metrology applications while retaining the
355 hallmark accuracy of comb-based metrology.

356

357 **ACKNOWLEDGMENTS**

358 We acknowledge Jennifer Ellis, Tara Fortier, Kevin Cossel, William Swann, Benjamin Stuhl,
359 and Brian Washburn for helpful discussions.

360

361 **AUTHOR CONTRIBUTIONS**

362 All authors contributed extensively to this work.

363

364 **COMPETING INTERESTS**

365 The authors declare no competing interests.

366

367 **DATA AVAILABILITY**

368 The data for the figures in this manuscript are available for download at
369 <https://data.nist.gov/sdp/><to be provided>.

370

371 **CODE AVAILABILITY**

372 The mathematics and algorithms necessary to create a time programmable frequency comb are
373 described between the main text and the methods section.

374

375 **MATERIALS AND CORRESPONDANCE**

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377 (nathan.newbury@nist.gov).

378

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